## Low Voltage, 1.15 V to 5.5 V, 4-Channel, Bidirectional Logic Level Translator

## Data Sheet

## FEATURES

## Bidirectional level translation <br> Operates from 1.15 V to 5.5 V <br> Low quiescent current < $5 \boldsymbol{\mu A}$ <br> No direction pin <br> Qualified for automotive applications

## APPLICATIONS

SPI ${ }^{\oplus}$, MICROWIRE ${ }^{\text {TM }}$ level translation
Low voltage ASIC level translation
Smart card readers
Cell phones and cell phone cradles
Portable communications devices
Telecommunications equipment
Network switches and routers
Storage systems (SAN/NAS)
Computing/server applications
GPS
Portable POS systems
Low cost serial interfaces

## GENERAL DESCRIPTION

The ADG3304 is a bidirectional logic level translator that contains four bidirectional channels. It can be used in multivoltage digital system applications, such as data transfer, between a low voltage digital signal processing controller and a higher voltage device using SPI and MICROWIRE interfaces. The internal architecture allows the device to perform bidirectional logic level translation without an additional signal to set the direction in which the translation takes place.

The voltage applied to $\mathrm{V}_{\mathrm{CCA}}$ sets the logic levels on the A side of the device, while $V_{C C Y}$ sets the levels on the $Y$ side. For proper operation, Vcca must always be less than Vccy. The Vcca-compatible logic signals applied to the A side of the device appear as $\mathrm{V}_{\mathrm{CCY}}$-compatible levels on the Y side. Similarly, $\mathrm{V}_{\mathrm{CCY}}$-compatible logic levels applied to the Y side of the device appear as $\mathrm{V}_{\mathrm{CCA}}{ }^{-}$ compatible logic levels on the A side.

FUNCTIONAL BLOCK DIAGRAM


Figure 1.

The enable pin (EN) provides three-state operation on both the A side and the Y side pins. When the EN pin is pulled low, the terminals on both sides of the device are in the high impedance state. The EN pin is referred to the $\mathrm{V}_{\mathrm{CCA}}$ supply voltage and driven high for normal operation.

The ADG3304 is available in compact 14-lead TSSOP, 12-ball WLCSP, and 20-lead LFCSP. It is guaranteed to operate over the 1.15 V to 5.5 V supply voltage range.

## PRODUCT HIGHLIGHTS

1. Bidirectional level translation.
2. Fully guaranteed over the 1.15 V to 5.5 V supply range.
3. No direction pin.
4. Available in 14-lead TSSOP, 12-ball WLCSP, and 20-lead LFCSP.

Rev. D

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ADG3304

## SPECIFICATIONS

$\mathrm{V}_{C C Y}=1.65 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{C C A}=1.15 \mathrm{~V}$ to $\mathrm{V}_{\mathrm{CCY}}, \mathrm{GND}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$. All specifications $\mathrm{T}_{\mathrm{MIN}}$ to $\mathrm{T}_{\mathrm{MAX}}$, unless otherwise noted.
Table 1.


| Parameter | Symbol | Test Conditions/Comments | B Version ${ }^{1}$ |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max |  |
| SWITCHING CHARACTERISTICS² |  |  |  |  |  |  |
| $3.3 \mathrm{~V} \pm 0.3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CCA}} \leq \mathrm{V}_{C C Y}, \mathrm{~V}_{C C Y}=5 \mathrm{~V} \pm 0.5 \mathrm{~V}$ |  |  |  |  |  |  |
| $\mathrm{A} \rightarrow \mathrm{Y}$ Level Translation |  | $R_{S}=R_{T}=50 \Omega, C_{L}=50 \mathrm{pF}$, see Figure 37 |  |  |  |  |
| Propagation Delay | $t_{P, ~ A \rightarrow Y}$ |  |  | 6 | 10 | ns |
| Rise Time | $\mathrm{t}_{\mathrm{R}, A \rightarrow Y}$ |  |  | 2 | 3.5 | ns |
| Fall Time | $t_{F, A \rightarrow Y}$ |  |  | 2 | 3.5 | ns |
| Maximum Data Rate | $D_{\text {MAX }, ~}^{\text {a }}$, ${ }^{\text {r }}$ |  | 50 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SKEW, }}$, $\rightarrow$ Y |  |  | 2 | 4 | ns |
| Part-to-Part Skew | $\mathrm{t}_{\text {PPSKEW, }} \mathrm{A} \rightarrow \mathrm{Y}$ |  |  |  | 3 | ns |
| $Y \rightarrow$ A Level Translation |  | $R_{S}=R_{T}=50 \Omega, C_{L}=15 \mathrm{pF}$, see Figure 38 |  |  |  |  |
| Propagation Delay | $t_{P, Y \rightarrow A}$ |  |  | 4 | 7 | ns |
| Rise Time | $t_{R, Y \rightarrow A}$ |  |  | 1 | 3 | ns |
| Fall Time | $t_{F, Y \rightarrow A}$ |  |  | 3 | 7 | ns |
| Maximum Data Rate | $\mathrm{D}_{\text {MAX }, Y \rightarrow \mathrm{~A}}$ |  | 50 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SKEW, }} \mathrm{Y} \rightarrow \mathrm{A}$ |  |  | 2 | 3.5 | ns |
| Part-to-Part Skew | $\mathrm{t}_{\text {PPSKEW, }} \mathrm{Y} \rightarrow \mathrm{A}$ |  |  |  | 2 | ns |
| $1.8 \mathrm{~V} \pm 0.15 \mathrm{~V} \leq \mathrm{V}_{C C A} \leq \mathrm{V}_{C C Y}, \mathrm{~V}_{C C Y}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ |  |  |  |  |  |  |
| $A \rightarrow Y$ Translation |  | $R_{S}=R_{T}=50 \Omega, C_{L}=50 \mathrm{pF}$, see Figure 37 |  |  |  |  |
| Propagation Delay | $t_{P, A \rightarrow Y}$ |  |  | 8 | 11 | ns |
| Rise Time | $t_{R, A \rightarrow Y}$ |  |  | 2 | 5 | ns |
| Fall Time | $t_{F, A \rightarrow Y}$ |  |  | 2 | 5 | ns |
| Maximum Data Rate | $\mathrm{D}_{\text {MAX, }} \rightarrow$, ${ }^{\text {r }}$ |  | 50 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SKEW, }} \mathrm{A}_{\text {, }}$ |  |  | 2 | 4 | ns |
| Part-to-Part Skew | t PPSKEW, $A \rightarrow Y$ |  |  |  | 4 | ns |
| $\mathrm{Y} \rightarrow \mathrm{A}$ Translation |  | $\mathrm{R}_{S}=\mathrm{R}_{T}=50 \Omega, C_{L}=15 \mathrm{pF}$, see Figure 38 |  |  |  |  |
| Propagation Delay | $t_{P, Y \rightarrow A}$ |  |  | 5 | 8 | ns |
| Rise Time | $t_{R, Y \rightarrow A}$ |  |  | 2 | 3.5 | ns |
| Fall Time | $t_{F, Y \rightarrow A}$ |  |  | 2 | 3.5 | ns |
| Maximum Data Rate | $\mathrm{D}_{\text {MAX }, ~}^{\text {Y } \rightarrow \text { A }}$ |  | 50 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SKEW, }} \mathrm{Y} \rightarrow \mathrm{A}$ |  |  | 2 | 3 | ns |
| Part-to-Part Skew | $\mathrm{t}_{\text {PPSKEW, }} \mathrm{Y} \rightarrow \mathrm{A}$ |  |  |  | 3 | ns |
| 1.15 V to 1.3 $\mathrm{V} \leq \mathrm{V}_{C C A} \leq \mathrm{V}_{C C Y}, \mathrm{~V}_{C C Y}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ |  |  |  |  |  |  |
| $\mathrm{A} \rightarrow$ Y Translation |  | $R_{S}=R_{T}=50 \Omega, C_{L}=50 \mathrm{pF}$, see Figure 37 |  |  |  |  |
| Propagation Delay | $t_{P, A \rightarrow Y}$ |  |  | 9 | 18 | ns |
| Rise Time | $\mathrm{t}_{\mathrm{R}, \mathrm{A} \rightarrow \mathrm{Y}}$ |  |  | 3 | 5 | ns |
| Fall Time | $t_{F, A \rightarrow Y}$ |  |  | 2 | 5 | ns |
| Maximum Data Rate | $D_{\text {MAX }, ~ A \rightarrow Y ~}^{\prime}$ |  | 40 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SKEW, }{ }_{\text {a }} \text {, }}$ |  |  | 2 | 5 | ns |
| Part-to-Part Skew | $\mathrm{t}_{\text {PPSKEW, }} \mathrm{A} \rightarrow \mathrm{Y}$ |  |  |  | 10 | ns |
| $\mathrm{Y} \rightarrow \mathrm{A}$ Translation |  | $R_{S}=R_{T}=50 \Omega, C_{L}=15 \mathrm{pF}$, see Figure 38 |  |  |  |  |
| Propagation Delay | $t_{P, Y \rightarrow A}$ |  |  | 5 | 9 | ns |
| Rise Time | $\mathrm{t}_{\mathrm{R}, \mathrm{Y} \rightarrow \mathrm{A}}$ |  |  | 2 | 4 | ns |
| Fall Time | $t_{F, Y \rightarrow A}$ |  |  | 2 | 4 | ns |
| Maximum Data Rate | $\mathrm{D}_{\text {MAX, }, ~}^{\text {¢ }}$ A |  | 40 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SkEw, }} \mathrm{Y} \rightarrow \mathrm{A}$ |  |  | 2 | 4 | ns |
| Part-to-Part Skew | $\mathrm{t}_{\text {PPSKEW, }} \mathrm{Y} \rightarrow \mathrm{A}$ |  |  |  | 4 | ns |


| Parameter | Symbol | Test Conditions/Comments | B Version ${ }^{1}$ |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max |  |
| 1.15 V to 1.3 $\mathrm{V} \leq \mathrm{V}_{C \subset A} \leq \mathrm{V}_{C C Y}, \mathrm{~V}_{C C Y}=1.8 \mathrm{~V} \pm 0.3 \mathrm{~V}$ |  |  |  |  |  |  |
| $A \rightarrow Y$ Translation |  | $\mathrm{R}_{S}=\mathrm{R}_{T}=50 \Omega, \mathrm{C}_{L}=50 \mathrm{pF}$, see Figure 37 |  |  |  |  |
| Propagation Delay | $t_{P, A \rightarrow Y}$ |  |  | 12 | 25 | ns |
| Rise Time | $t_{R, A \rightarrow Y}$ |  |  | 7 | 12 | ns |
| Fall Time | $\mathrm{t}_{\mathrm{F}, \mathrm{A} \rightarrow \mathrm{Y}}$ |  |  | 3 | 5 | ns |
| Maximum Data Rate | $D_{\text {MAX }, ~}^{\text {a }}$, ${ }^{\text {r }}$ |  | 25 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SKEW, }}$ A $\rightarrow$ Y |  |  | 2 | 5 | ns |
| Part-to-Part Skew | $\mathrm{t}_{\text {PPSKEW, }}$ A $\rightarrow$ Y |  |  |  | 15 | ns |
| $\mathrm{Y} \rightarrow \mathrm{A}$ Translation |  | $\mathrm{R}_{\mathrm{s}}=\mathrm{R}_{T}=50 \Omega, \mathrm{C}_{L}=15 \mathrm{pF}$, see Figure 38 |  |  |  |  |
| Propagation Delay | $t_{P, Y \rightarrow A}$ |  |  | 14 | 35 | ns |
| Rise Time | $t_{R, Y \rightarrow A}$ |  |  | 5 | 16 | ns |
| Fall Time | $t_{F, Y \rightarrow A}$ |  |  | 2.5 | 6.5 | ns |
| Maximum Data Rate | $D_{M A X, Y \rightarrow A}$ |  | 25 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SKEW, }} \rightarrow \mathrm{A}$ |  |  | 3 | 6.5 | ns |
| Part-to-Part Skew | $\mathrm{t}_{\text {PPSKEW, }} \mathrm{Y} \rightarrow \mathrm{A}$ |  |  |  | 23.5 | ns |
| $\begin{aligned} & 2.5 \mathrm{~V} \pm 0.2 \mathrm{~V} \leq \mathrm{V}_{\mathrm{ccA}} \leq \mathrm{V}_{\mathrm{cc}}, \mathrm{~V}_{\mathrm{CCY}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V} \\ & \mathrm{~A} \rightarrow \mathrm{Y} \text { Translation } \end{aligned}$ |  | $R_{S}=R_{T}=50 \Omega, C_{L}=50 \mathrm{pF}$, see Figure 37 |  |  |  |  |
| Propagation Delay | $t_{P, A \rightarrow Y}$ |  |  | 7 | 10 | ns |
| Rise Time | $t_{R, A \rightarrow Y}$ |  |  | 2.5 | 4 | ns |
| Fall Time |  |  |  | 2 | 5 | ns |
| Maximum Data Rate | $D_{\text {MAX, } A \rightarrow Y}$ |  | 60 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\text {SKEW, }} \rightarrow \boldsymbol{A} \boldsymbol{Y}$ |  |  | 1.5 | 2 | ns |
| Part-to-Part Skew | $t_{\text {PPSKEW, }}$ A $\rightarrow$ Y |  |  |  | 4 | ns |
| $\mathrm{Y} \rightarrow \mathrm{A}$ Translation |  | $\mathrm{R}_{S}=\mathrm{R}_{T}=50 \Omega, \mathrm{C}_{L}=15 \mathrm{pF}$, see Figure 38 |  |  |  |  |
| Propagation Delay | $t_{P, Y \rightarrow A}$ |  |  | 5 | 8 | ns |
| Rise Time | $t_{R, Y \rightarrow A}$ |  |  | 1 | 4 | ns |
| Fall Time | $t_{f, Y \rightarrow A}$ |  |  | 3 | 5 | ns |
| Maximum Data Rate | $D_{\text {MAX }, Y \rightarrow A}$ |  | 60 |  |  | Mbps |
| Channel-to-Channel Skew | $\mathrm{t}_{\mathrm{SKEW}, \mathrm{Y} \rightarrow \mathrm{~A}}$ |  |  | 2 | $3$ | ns |
| Part-to-Part Skew | $\mathrm{tPPSKEW}, \mathrm{Y} \rightarrow \mathrm{A}$ |  |  |  | 3 | ns |
| POWER REQUIREMENTS |  |  |  |  |  |  |
| Power Supply Voltages | V cca | $\mathrm{V}_{\text {cCA }} \leq \mathrm{V}_{\text {cCY }}$ | 1.15 |  | 5.5 | V |
|  | $\mathrm{V}_{\text {CCY }}$ |  | 1.65 |  | 5.5 | V |
| Quiescent Power Supply Current | Icca | $\begin{aligned} & \mathrm{V}_{\mathrm{A}}=0 \mathrm{~V} / \mathrm{V}_{\mathrm{CCA}}, \mathrm{~V}_{\mathrm{Y}}=0 \mathrm{~V} / \mathrm{V}_{\mathrm{CCY}}, \\ & \mathrm{~V}_{\mathrm{CCA}}=\mathrm{V}_{\mathrm{CCY}}=5.5 \mathrm{~V}, \mathrm{EN}=1 \end{aligned}$ |  | 0.17 | 5 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{\text {CCY }}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{A}}=0 \mathrm{~V} / \mathrm{V}_{\mathrm{CCA}}, \mathrm{~V}_{\mathrm{Y}}=0 \mathrm{~V} / \mathrm{V}_{\mathrm{CCY}}, \\ & \mathrm{~V}_{\mathrm{CCA}}=\mathrm{V}_{\mathrm{CCY}}=5.5 \mathrm{~V}, \mathrm{EN}=1 \end{aligned}$ |  | 0.27 | 5 | $\mu \mathrm{A}$ |
| Three-State Mode Power Supply Current | $\mathrm{I}_{\mathrm{Hi}-\mathrm{Z}, \mathrm{A}}$ | $\mathrm{V}_{C C A}=\mathrm{V}_{\text {CCY }}=5.5 \mathrm{~V}, \mathrm{EN}=0$ |  | 0.1 | 5 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{\text {Hi-Z, Y }}$ | $\mathrm{V}_{C C A}=\mathrm{V}_{C C Y}=5.5 \mathrm{~V}$, EN $=0$ |  | 0.1 | 5 | $\mu \mathrm{A}$ |

[^0]
## ABSOLUTE MAXIMUM RATINGS

$\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise noted.
Table 2.

| Parameter | Rating |
| :---: | :---: |
| Vcca to GND | -0.3 V to +7V |
| Vccr to GND | V cca to +7 V |
| Digital Inputs (A) | -0.3 V to ( $\left.\mathrm{V}_{\text {cca }}+0.3 \mathrm{~V}\right)$ |
| Digital Inputs ( Y ) | -0.3 V to ( $\left.\mathrm{V}_{\text {ccr }}+0.3 \mathrm{~V}\right)$ |
| EN to GND | -0.3 V to +7 V |
| Operating Temperature Range | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Junction Temperature | $150^{\circ} \mathrm{C}$ |
| $\theta_{\mathrm{JA}}$ Thermal Impedance (4-Layer Board) |  |
| 14-Lead TSSOP | $89.21^{\circ} \mathrm{C} / \mathrm{W}$ |
| 12-Ball WLCSP | $120^{\circ} \mathrm{C} / \mathrm{W}$ |
| 20-Lead LFCSP | $30.4^{\circ} \mathrm{C} / \mathrm{W}$ |
| Lead Temperature, Soldering | As per JEDEC J-STD-020 |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those listed in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Only one absolute maximum rating can be applied at any one time.

## ESD CAUTION

|  | ESD (electrostatic discharge) sensitive device. <br> Charged devices and circuit boards can discharge <br> without detection. Although this product features <br> patented or proprietary protection circuitry, damage <br> may occur on devices subjected to high energy ESD. <br> Therefore, proper ESD precautions should be taken to <br> avoid performance degradation or loss of functionality. |
| :--- | :--- |

## PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS



Figure 2. 14-Lead TSSOP Pin Configuration



NC = NO CONNECT
THE EXPOSED PADDLE CAN BE TIED TO GND OR LEFT FLOATING. DO NOT TIE IT TO $\mathrm{V}_{\text {CCA }}$ or $\mathrm{V}_{\mathrm{CcY}}$

Figure 4. 20-Lead LFCSP_VQ
Pin Configuration

Table 3. 14-Lead TSSOP and 20-lead LFCSP Pin Function Descriptions

| Pin No. |  | Mnemonic | Description |
| :---: | :---: | :---: | :---: |
| TSSOP | LFCSP |  |  |
| 1 | 19 | $\mathrm{V}_{\text {CCA }}$ | Power Supply Voltage Input for the A 1 to $\mathrm{A} 4 \mathrm{I} / \mathrm{O}$ Pins ( $1.15 \mathrm{~V} \leq \mathrm{V}_{\text {ccA }} \leq \mathrm{V}_{\text {cç }}$ ). |
| 2 | 20 | A1 | Input/Output A 1 . Referenced to V cca . |
| 3 | 2 | A2 | Input/Output $A 2$. Referenced to $V_{\text {cca }}$. |
| 4 | 3 | A3 | Input/Output $A 3$. Referenced to $V_{\text {cca }}$. |
| 5 | 4 | A4 | Input/Output A4. Referenced to V cca . |
| 6,9 | 1, 5, 6, 7, 10, 11, 15, 16 | NC | No Connect. |
| 7 | 8 | GND | Ground. |
| 8 | 9 | EN | Active High Enable Input. |
| 10 | 12 | Y4 | Input/Output Y4. Referenced to V ccr . |
| 11 | 13 | Y3 | Input/Output Y3. Referenced to $\mathrm{V}_{\text {ccr }}$. |
| 12 | 14 | Y2 | Input/Output Y . Referenced to $\mathrm{V}_{\text {ccr }}$. |
| 13 | 17 | Y1 | Input/Output Y . Referenced to $\mathrm{V}_{\text {ccr }}$. |
| 14 | 18 | $\mathrm{V}_{\mathrm{CCY}}$ | Power Supply Voltage Input for the Y 1 to $\mathrm{Y} 4 \mathrm{I} / \mathrm{O}$ Pins $\left(1.65 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CcY}} \leq 5.5 \mathrm{~V}\right)$. |

Table 4. 12-Ball WLCSP Pin Function Descriptions

| Bump No. | Mnemonic | Description |
| :---: | :---: | :---: |
| A1 | Y1 | Input/Output Y1. Referenced to $\mathrm{V}_{\text {ccr }}$. |
| B1 | Y2 | Input/Output Y 2. Referenced to $\mathrm{V}_{\text {ccr }}$. |
| C1 | Y3 | Input/Output Y3. Referenced to $\mathrm{V}_{\text {ccr }}$. |
| D1 | Y4 | Input/Output Y4. Referenced to $\mathrm{V}_{\text {ccr }}$. |
| A2 | $\mathrm{V}_{\text {cCr }}$ | Power Supply Voltage Input for the Y 1 to $\mathrm{Y} 4 \mathrm{I} / \mathrm{O}$ Pins $\left(1.65 \mathrm{~V} \leq \mathrm{V}_{\mathrm{ccY}} \leq 5.5 \mathrm{~V}\right)$. |
| B2 | Vcca | Power Supply Voltage Input for the A1 to A4 I/O Pins ( $1.15 \mathrm{~V} \leq \mathrm{V}_{\text {cca }} \leq \mathrm{V}_{\text {ccr }}$ ). |
| C2 | EN | Active High Enable Input. |
| D2 | GND | Ground. |
| A3 | A1 | Input/Output A1. Referenced to $\mathrm{V}_{\text {cca }}$. |
| B3 | A2 | Input/Output A2. Referenced to V cca . |
| C3 | A3 | Input/Output A3. Referenced to V cca . |
| D3 | A4 | Input/Output A4. Referenced to $\mathrm{V}_{\text {cca }}$. |

## TYPICAL PERFORMANCE CHARACTERISTICS



Figure 5. Icca vs. Data Rate ( $A \rightarrow Y$ Level Translation)


Figure 6. Iccr vs. Data Rate ( $A \rightarrow Y$ Level Translation)


Figure 7. IccA vs. Data Rate ( $Y \rightarrow$ A Level Translation)


Figure 8. Iccy vs. Data Rate $(Y \rightarrow$ A Level Translation)


Figure 9. Iccy vs. Capacitive Load at Pin $Y$ for $A \rightarrow Y(1.2 V \rightarrow 1.8 V)$ Level Translation


Figure 10. Icca Vs. Capacitive Load at Pin $A$ for $Y \rightarrow A(1.8 \mathrm{~V} \rightarrow 1.2 \mathrm{~V})$ Level Translation


Figure 11. Iccy vs. Capacitive Load at Pin $Y$ for $A \rightarrow Y(1.8 V \rightarrow 3.3 V)$ Level Translation


Figure 12. IccA vs. Capacitive Load at Pin $A$ for $Y \rightarrow A(3.3 \mathrm{~V} \rightarrow 1.8 \mathrm{~V})$ Level Translation


Figure 13. Iccy vs. Capacitive Load at Pin $Y$ for $A \rightarrow Y(3.3 V \rightarrow 5 V)$ Level Translation


Figure 14. Icca Vs. Capacitive Load at Pin A for $Y \rightarrow A(5 \mathrm{~V} \rightarrow 3.3 \mathrm{~V})$ Level Translation


Figure 15. Rise Time vs. Capacitive Load at Pin $Y$ ( $A \rightarrow Y$ Level Translation)


Figure 16. Fall Time vs. Capacitive Load at Pin $Y(A \rightarrow Y$ Level Translation)


Figure 17. Rise Time vs. Capacitive Load at Pin $A(Y \rightarrow A$ Level Translation)


Figure 18. Fall Time vs. Capacitive Load at Pin A $(Y \rightarrow$ A Level Translation)


Figure 19. Propagation Delay ( $t_{\text {PLH }}$ ) vs.
Capacitive Load at Pin $Y(A \rightarrow Y$ Level Translation)


Figure 20. Propagation Delay ( $t_{\text {PHL }}$ ) vs. Capacitive Load at Pin $Y$ ( $A \rightarrow Y$ Level Translation)


Figure 21. Propagation Delay ( $t_{P L H}$ ) vs. Capacitive Load at Pin A $(Y \rightarrow A$ Level Translation)


Figure 22. Propagation Delay ( $t_{\text {PHL }}$ ) vs.
Capacitive Load at Pin $A(Y \rightarrow A$ Level Translation)


Figure 23. Eye Diagram at $Y$ Output (1.2 V to 1.8 V Level Translation, 25 Mbps )


Figure 24. Eye Diagram at A Output (1.8 V to 1.2 V Level Translation, 25 Mbps )


Figure 25. Eye Diagram at $Y$ Output (1.8 V to 3.3 V Level Translation, 50 Mbps )


Figure 26. Eye Diagram at A Output (3.3 V to 1.8 V Level Translation, 50 Mbps )


Figure 27. Eye Diagram at $Y$ Output (3.3 V to 5 V Level Translation, 50 Mbps )


Figure 28. Eye Diagram at A Output (5 V to 3.3 V Level Translation, 50 Mbps )

## TEST CIRCUITS



Figure 29. $V_{O H} / V_{O L}$ Voltages at Pin $A$


Figure 30. $V_{\text {OH }} / V_{o L}$ Voltages at Pin $Y$


Figure 31. Three-State Leakage Current at Pin A


Figure 32. Three-State Leakage Current at Pin $Y$


Figure 33. EN Pin Leakage Current


Figure 34. Capacitance at Pin A


Figure 35. Capacitance at Pin $Y$


Figure 36. Enable Time


Figure 37. Switching Characteristics ( $A \rightarrow Y$ Level Translation)

## TERMINOLOGY

$\mathrm{V}_{\mathrm{IHA}}$
Logic input high voltage at Pin A1 to Pin A4.
$V_{\text {ILA }}$
Logic input low voltage at Pin A1 to Pin A4.
Voha
Logic output high voltage at Pin A1 to Pin A4.

## Vola

Logic output low voltage at Pin A1 to Pin A4.
$\mathrm{C}_{\mathrm{A}}$
Capacitance measured at Pin A1 to Pin A4 $(\mathrm{EN}=0)$.

## $\mathbf{I}_{\text {LA, Hi-Z }}$

Leakage current at Pin A1 to Pin A4 when EN = 0 (high impedance state at Pin A1 to Pin A4).
$\mathrm{V}_{\mathrm{IHY}}$
Logic input high voltage at Pin Y1 to Pin Y4.
$V_{\text {IIY }}$
Logic input low voltage at Pin Y1 to Pin Y4.
Voнy
Logic output high voltage at Pin Y1 to Pin Y4.

## Voly

Logic output low voltage at Pin Y1 to Pin Y4.
$\mathrm{C}_{\mathrm{Y}}$
Capacitance measured at Pin Y1 to Pin Y4 (EN = 0).

## $\mathbf{I}_{\text {LY, Hi-Z }}$

Leakage current at Pin Y1 to Pin Y4 when EN $=0$ (high
impedance state at Pin Y1 to Pin Y4).
$V_{\text {ihen }}$
Logic input high voltage at the EN pin.
Vilen
Logic input low voltage at the EN pin.
$\mathrm{C}_{\text {EN }}$
Capacitance measured at EN pin.
I Len
Enable (EN) pin leakage current.
ten
Three-state enable time for Pin A1 to Pin A4 and Pin Y1 to Pin Y4.
$\mathbf{t}_{\mathrm{P}, \mathrm{A} \rightarrow \mathrm{Y}}$
Propagation delay when translating logic levels in the $A \rightarrow Y$ direction.
$\mathbf{t}_{\mathrm{R}, \mathrm{A} \rightarrow \mathrm{Y}}$
Rise time when translating logic levels in the $A \rightarrow Y$ direction.
$\mathrm{T}_{\mathrm{E}, \mathrm{A} \rightarrow \mathrm{Y}}$
Fall time when translating logic levels in the $\mathrm{A} \rightarrow \mathrm{Y}$ direction.
$\mathrm{D}_{\mathrm{MAX}, \mathrm{A} \rightarrow \mathrm{Y}}$
Guaranteed data rate when translating logic levels in the $\mathrm{A} \rightarrow \mathrm{Y}$ direction under the driving and loading conditions specified in Table 1.
$\mathrm{T}_{\text {SKEW, } \mathrm{A} \rightarrow \mathrm{Y}}$
Difference between propagation delays on any two channels when translating logic levels in the $\mathrm{A} \rightarrow \mathrm{Y}$ direction.
$\mathbf{t}_{\text {pPSKEW, } A \rightarrow Y}$
Difference in propagation delay between any one channel and the same channel on a different part (under same driving/ loading conditions) when translating in the $\mathrm{A} \rightarrow \mathrm{Y}$ direction.
$\mathbf{t}_{\mathrm{P}, \mathrm{Y} \rightarrow \mathrm{A}}$
Propagation delay when translating logic levels in the $\mathrm{Y} \rightarrow \mathrm{A}$ direction.
$\mathbf{t}_{\mathrm{R}, \mathrm{Y} \rightarrow \mathrm{A}}$
Rise time when translating logic levels in the $\mathrm{Y} \rightarrow \mathrm{A}$ direction.
$\mathbf{t}_{\mathrm{F}, \mathrm{Y} \rightarrow \mathrm{A}}$
Fall time when translating logic levels in the $\mathrm{Y} \rightarrow \mathrm{A}$ direction.
$\mathrm{D}_{\mathrm{MAX}, \mathrm{Y} \rightarrow \mathrm{A}}$
Guaranteed data rate when translating logic levels in the $\mathrm{Y} \rightarrow \mathrm{A}$ direction under the driving and loading conditions specified in Table 1.
$\mathbf{t s k e w , ~}_{\text {Y } \rightarrow \mathrm{A}}$
Difference between propagation delays on any two channels when translating logic levels in the $\mathrm{Y} \rightarrow \mathrm{A}$ direction.
tppskew, $\mathrm{Y} \rightarrow \mathrm{A}$
Difference in propagation delay between any one channel and the same channel on a different part (under the same driving/ loading conditions) when translating in the $\mathrm{Y} \rightarrow \mathrm{A}$ direction.

Vcca
$\mathrm{V}_{\mathrm{CCA}}$ supply voltage.
$V_{\text {CCy }}$
$\mathrm{V}_{C C Y}$ supply voltage.
$I_{\text {CCA }}$
$\mathrm{V}_{\mathrm{CCA}}$ supply current.
I CCy $^{\prime}$
$\mathrm{V}_{\mathrm{CCY}}$ supply current.
$\mathbf{I}_{\mathrm{Hi}-\mathrm{Z}, \mathrm{A}}$
VCCA supply current during three-state mode ( $\mathrm{EN}=0$ ).
$\mathbf{I}_{\mathrm{Hi}-\mathrm{Z}, \mathrm{Y}}$
$\mathrm{V}_{\mathrm{CCY}}$ supply current during three-state mode $(\mathrm{EN}=0)$.

## THEORY OF OPERATION

The ADG3304 level translator allows the level shifting necessary for data transfer in a system where multiple supply voltages are used. The device requires two supplies, $\mathrm{V}_{\text {CCA }}$ and $\mathrm{V}_{\mathrm{CCY}}\left(\mathrm{V}_{\mathrm{CCA}} \leq \mathrm{V}_{\mathrm{CCY}}\right)$. These supplies set the logic levels on each side of the device. When driving the A pins, the device translates the $\mathrm{V}_{\text {CCA }}$-compatible logic levels to $\mathrm{V}_{\mathrm{CCY}}$-compatible logic levels available at the Y pins. Similarly, because the device is capable of bidirectional translation, when driving the Y pins, the $\mathrm{V}_{\mathrm{CCY}}-$ compatible logic levels are translated to $\mathrm{V}_{\mathrm{CCA}}$-compatible logic levels available at the A pins. When $\mathrm{EN}=0$, Pin A1 to Pin A4 and Pin Y1 to Pin Y4 are three-stated. When EN is driven high, the ADG3304 goes into normal operation mode and performs level translation.

## LEVEL TRANSLATOR ARCHITECTURE

The ADG3304 consists of four bidirectional channels. Each channel can translate logic levels in either the $\mathrm{A} \rightarrow \mathrm{Y}$ or the $\mathrm{Y} \rightarrow \mathrm{A}$ direction. It uses a one-shot accelerator architecture, which ensures excellent switching characteristics. Figure 39 shows a simplified block diagram of a bidirectional channel.


The logic level translation in the $A \rightarrow Y$ direction is performed using a level translator (U1) and an inverter (U2), while the translation in the $\mathrm{Y} \rightarrow \mathrm{A}$ direction is performed using Inverter U3 and Inverter U4. The one-shot generator detects a rising or falling edge present on either the A side or the Y side of the channel. It sends a short pulse that turns on the PMOS transistors (T1 to T2) for a rising edge, or the NMOS transistors (T3 to T4) for a falling edge. This charges/discharges the capacitive load faster, which results in faster rise and fall times.

The inputs of the unused channels (A or Y) should be tied to their corresponding $V_{C C}$ rail ( $V_{C C A}$ or $V_{C C Y}$ ) or to GND.

## INPUT DRIVING REQUIREMENTS

To ensure correct operation of the ADG3304, the circuit that drives the input of the ADG3304 channels should have an output impedance of less than or equal to $150 \Omega$ and a minimum peak current driving capability of 36 mA .

## OUTPUT LOAD REQUIREMENTS

The ADG3304 level translator is designed to drive CMOScompatible loads. If current-driving capability is required, it is recommended to use buffers between the ADG3304 outputs and the load.

## ENABLE OPERATION

The ADG3304 provides three-state operation at the A and Y I/O pins by using the enable pin (EN), as shown in Table 5.
Table 5. Truth Table

| EN | Y I/O Pins | A I/O Pins |
| :--- | :--- | :--- |
| 0 | $\mathrm{Hi}^{1}$ | $\mathrm{Hi}^{1} \mathrm{Z}^{1}$ |
| 1 | Normal operation $^{2}$ | Normal operation $^{2}$ |

${ }^{1}$ High impedance state.
${ }^{2}$ In normal operation, the ADG3304 performs level translation.
While EN $=0$, the ADG3304 enters into three-state mode. In this mode, the current consumption from both the $V_{C C A}$ and $V_{C C Y}$ supplies is reduced, allowing the user to save power, which is critical, especially on battery-operated systems. The EN input pin can be driven with either $\mathrm{V}_{\text {CCA }}$-compatible or $\mathrm{V}_{\text {CCY }}$-compatible logic levels.

## POWER SUPPLIES

For proper operation of the ADG3304, the voltage applied to the Vcca must be less than or equal to the voltage applied to $\mathrm{V}_{\mathrm{Ccy}}$. To meet this condition, the recommended power-up sequence is $\mathrm{V}_{\mathrm{CCY}}$ first and then $\mathrm{V}_{\mathrm{CCA}}$. The ADG3304 operates properly only after both supply voltages reach their nominal values. It is not recommended to use the part in a system where, during power-up, $\mathrm{V}_{\mathrm{CCA}}$ can be greater than $\mathrm{V}_{\mathrm{CCY}}$ due to a significant increase in the current taken from the $\mathrm{V}_{\text {CCA }}$ supply. For optimum performance, the $V_{\text {CCa }}$ pin and $V_{\text {ccy }}$ pin should be decoupled to GND as close as possible to the device.

## DATA RATE

The maximum data rate at which the device is guaranteed to operate is a function of the $\mathrm{V}_{\mathrm{CCA}}$ and $\mathrm{V}_{\mathrm{CCY}}$ supply voltage combination and the load capacitance. It is given by the maximum frequency of a square wave that can be applied to the device, which meets the $\mathrm{V}_{\text {OH }}$ and $V_{\text {oL }}$ levels at the output and does not exceed the maximum junction temperature (see the Absolute Maximum Ratings section). Table 6 shows the guaranteed data rates at which the ADG3304 can operate in both directions ( $\mathrm{A} \rightarrow \mathrm{Y}$ or $\mathrm{Y} \rightarrow \mathrm{A}$ level translation) for various $\mathrm{V}_{\mathrm{CCA}}$ and $V_{C C Y}$ supply combinations.

Table 6. Guaranteed Data Rate (Mbps) ${ }^{1}$

| Vcca | $\mathrm{V}_{\mathrm{ccr}}$ |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | $\begin{gathered} 1.8 \mathrm{~V} \\ \text { (1.65 V to } 1.95 \mathrm{~V} \text { ) } \end{gathered}$ | $\begin{gathered} 2.5 \mathrm{~V} \\ (2.3 \mathrm{~V} \text { to } 2.7 \mathrm{~V}) \end{gathered}$ | $\begin{gathered} 3.3 \mathrm{~V} \\ (3.0 \mathrm{~V} \text { to } 3.6 \mathrm{~V}) \end{gathered}$ | $\begin{gathered} 5 \mathrm{~V} \\ (4.5 \mathrm{~V} \text { to } 5.5 \mathrm{~V}) \end{gathered}$ |
| 1.2 V (1.15 V to 1.3 V ) | 25 | 30 | 40 | 40 |
| 1.8 V (1.65 V to 1.95 V ) | - | 45 | 50 | 50 |
| 2.5 V (2.3V to 2.7 V ) | - | - | 60 | 50 |
| 3.3 V (3.0 V to 3.6 V ) | - | - | - | 50 |
| $5 \mathrm{~V}(4.5 \mathrm{~V}$ to 5.5 V$)$ | - | - | - | - |

[^1]
## APPLICATIONS

The ADG3304 is designed for digital circuits that operate at different supply voltages; therefore, logic level translation is required. The lower voltage logic signals are connected to the A pins, and the higher voltage logic signals are connected to the Y pins. The ADG3304 can provide level translation in both directions from $\mathrm{A} \rightarrow \mathrm{Y}$ or $\mathrm{Y} \rightarrow \mathrm{A}$ on all four channels, eliminating the need for a level translator IC for each direction. The internal architecture allows the ADG3304 to perform bidirectional level translation without an additional signal to set the direction in which the translation is made. It also allows simultaneous data flow in both directions on the same part, for example, when two channels translate in $\mathrm{A} \rightarrow \mathrm{Y}$ direction while the other two translate in $\mathrm{Y} \rightarrow \mathrm{A}$ direction. This simplifies the design by eliminating the timing requirements for the direction signal and reducing the number of ICs used for level translation.

Figure 40 shows an application where two microprocessors operating at 1.8 V and 3.3 V , respectively, can transfer data simultaneously using two full-duplex serial links, TX1/RX1 and TX2/RX2.


When the application requires level translation between a microprocessor and multiple peripheral devices, the ADG3304 I/O pins can be three-stated by setting $\mathrm{EN}=0$. This feature allows the ADG3304 to share the data buses with other devices without causing contention issues. Figure 41 shows an application where a 1.8 V microprocessor is connected to a 3.3 V peripheral device using the three-state feature.


## LAYOUT GUIDELINES

As with any high speed digital IC, the printed circuit board layout is important for the overall performance of the circuit. Care should be taken to ensure proper power supply bypass and return paths for the high speed signals. Each $\mathrm{V}_{\mathrm{CC}}$ pin ( $\mathrm{V}_{\mathrm{CCA}}$ and $\mathrm{V}_{\mathrm{CCY}}$ ) should be bypassed using low effective series resistance (ESR) and effective series inductance (ESI) capacitors placed as close as possible to the $V_{C C A}$ pin and the $V_{C C Y} p i n$. The parasitic inductance of the high speed signal track may cause significant overshoot. This effect can be reduced by keeping the length of the tracks as short as possible. A solid copper plane for the return path (GND) is also recommended.

## OUTLINE DIMENSIONS



Figure 42. 14-Lead Thin Shrink Small Outline Package [TSSOP]
(RU-14)
Dimensions shown in millimeters


Figure 43. 12-Ball Wafer Level Chip Scale Package [WLCSP] (CB-12-1)
Dimensions shown in millimeters


COMPLIANT TO JEDEC STANDARDS MO-220-VGGD-1
Figure 44. 20-Lead Lead Frame Chip Scale Package [LFCSP_VQ]
$4 \mathrm{~mm} \times 4 \mathrm{~mm}$ Body, Very Thin Quad
(CP-20-1)
Dimensions shown in millimeters

## ORDERING GUIDE

| Model $^{1,2}$ | Temperature Range | Package Description | Branding ${ }^{\mathbf{3}}$ | Package <br> Option |
| :--- | :--- | :--- | :--- | :--- |
| ADG3304BRUZ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 14-Lead Thin Shrink Small Outline Package [TSSOP] |  | RU-14 |
| ADG3304BRUZ-REEL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 14-Lead Thin Shrink Small Outline Package [TSSOP] |  | RU-14 |
| ADG3304BRUZ-REEL7 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 14-Lead Thin Shrink Small Outline Package [TSSOP] |  | RU-14 |
| ADG3304BCPZ-REEL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 20-Lead Lead Frame Chip Scale Package [LFCSP_VQ] |  | CP-20-1 |
| ADG3304BCPZ-REEL7 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 20-Lead Lead Frame Chip Scale Package [LFCSP_VQ] | CP-20-1 |  |
| ADG3304BCBZ-REEL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 12-Ball Wafer Level Chip Scale Package [WLCSP] | SDC | CB-12-1 |
| ADG3304BCBZ-REEL7 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 12-Ball Wafer Level Chip Scale Package [WLCSP] | SDC | CB-12-1 |
| ADG3304WBRUZ-REEL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 14-Lead Thin Shrink Small Outline Package [TSSOP] |  | RU-14 |

${ }^{1} \mathrm{Z}=$ RoHS Compliant Part.
${ }^{2} \mathrm{~W}=$ Qualified for Automotive Applications.
${ }^{3}$ Branding on these packages is limited to three characters due to space constraints.

## AUTOMOTIVE PRODUCTS

The ADG3304W model is available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that this automotive model may have specifications that differ from the commercial models; therefore, designers should review the Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.

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[^0]:    ${ }^{1} \mathrm{~T}_{\mathrm{A}}$ for typical specifications is $25^{\circ} \mathrm{C}$.
    ${ }^{2}$ Guaranteed by design, not production tested.

[^1]:    ${ }^{1}$ The load capacitance used is 50 pF when translating in the $\mathrm{A} \rightarrow \mathrm{Y}$ direction and 15 pF when translating in the $\mathrm{Y} \rightarrow \mathrm{A}$ direction.

